### 1.2 A, 30 V Step-Down DCIDC Converter

NO.EA-190-170609

## OUTLINE

The R1240x is a CMOS-based Step-down DC/DC converter with internal Nch high side Tr. ( $0.35 \Omega$ ), which can provide the maximum 1.2 A output current. The ICs consists of an Oscillator, a PWM control circuit, a Reference Voltage unit, an Error amplifier, phase compensation circuits, a slope circuit, a soft-start circuit, protection circuits, internal voltage regulators, and a switch for boot strap circuit. The ICs can make up a StepDown DC/DC Converter with the following external components: an inductor, resistors, a diode, and capacitors. The R1240x is a current mode operating type DC/DC converter which does not require external current sense resistor, and it works high speed response time, high efficiency and compatible with ceramic capacitors. Oscillator frequency is internally set at 1.25 MHz .
As a protection function, it has cycle by cycle peak current limit function, short protection function, thermal shutdown function and UVLO.

There are two types for short protection, A version has latch protection function with 2 ms delay time, and B version has fold-back protection function that keep operating at short condition with lower operating frequency and limiting the Lx current.

## FEATURES



- Internal Nch MOSFET Driver ............................................. Ron $=0.35 \Omega$
- Adjustable Output Voltage with External Resistor $\cdots 0.8 \mathrm{~V}$ to 15 V
- Feedback Voltage ................................................. $0.8 \mathrm{~V} \pm 1.5 \%$
- Peak Current Limit Function .......................................... A
- UVLO Function


- Ceramic Capacitor Compatible
- Stand-by Function ............................................................... $0 \mu \mathrm{~A}$



## APPLICATIONS

- Digital Home Appliances: Digital TVs, DVD Players
- OA Equipment: Printers, Fax
- Hand-held Communication Equipment, Cameras, VCRs, Camcorders
- Battery-powered Equipment

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## SELECTION GUIDE

In the R1240x, the Package, type of short protection (Latch or Fold-back) can be selected at the user's request.

## Selection Guide

| Product Name | Package | Quantity per Reel | Pb Free | Halogen Free |
| :---: | :---: | :---: | :---: | :---: |
| R1240K003*-TR | DFN(PLP)2527-10 | 5,000 pcs | Yes | Yes |
| R1240N001*-TR-FE | SOT-23-6W | 3,000 pcs | Yes | Yes |

*: Designation of Optional Function at off state are options as follows.
(A) Latch Type protection
(B) Fold-back Type protection

## BLOCK DIAGRAM



R1240x Block Diagram

## PIN DESCRIPTIONS



DFN(PLP)2527-10 Pin Configuration


SOT-23-6W Pin Configuration

R1240N001x Pin Description

| Pin No. | Symbol | Description |
| :---: | :---: | :--- |
| 1 | CE | Chip Enable Pin, Active with "H" |
| 2 | VIN | Power Supply Pin |
| 3 | Lx | Lx Switching Pin |
| 4 | BST | Bootstrap Pin |
| 5 | GND | Ground Pin |
| 6 | VFB | Feedback Pin |

## R1240K003x Pin Description

| Pin No. | Symbol |  |
| :---: | :---: | :--- |
| 1 | Lx | Lx Switching Pin |
| 2 | VIN | Power Supply Pin |
| 3 | VIN | Power Supply Pin |
| 4 | CE | Chip Enable Pin, Active with "H" |
| 5 | TEST | Test Pin (Open, do not connect to any line.) |
| 6 | GND | Ground Pin |
| 7 | NC | No Connection |
| 8 | VFB | Feedback Pin |
| 9 | NC | No Connection |
| 10 | BST | Bootstrap Pin |

Tab is GND level. (They are connected to the reverse side of this IC.) The tab is better to be connected to the GND, but leaving it open is also acceptable.

## ABSOLUTE MAXIMUM RATINGS



* Refer to Power Dissipation for detailed information.


## ABSOLUTE MAXIMUM RATINGS

Electronic and mechanical stress momentarily exceeded absolute maximum ratings may cause the permanent damages and may degrade the life time and safety for both device and system using the device in the field. The functional operation at or over these absolute maximum ratings is not assured.

## RECOMMENDED OPERATING CONDITIONS

Recommended Operating Conditions

| Symbol | Item | Rating | Unit |
| :---: | :--- | :---: | :---: |
| $\mathrm{V}_{\mathrm{IN}}$ | Operating Input Voltage | 4.5 to 30 | V |
| Ta | Operating Temperature Range | -40 to 85 | ${ }^{\circ} \mathrm{C}$ |

## RECOMMENDED OPERATING CONDITIONS

All of electronic equipment should be designed that the mounted semiconductor devices operate within the recommended operating conditions. The semiconductor devices cannot operate normally over the recommended operating conditions, even if when they are used over such conditions by momentary electronic noise or surge. And the semiconductor devices may receive serious damage when they continue to operate over the recommended operating conditions.

## ELECTRICAL CHARACTERISTICS

Electrical Characteristics
(Otherwise notified, $\mathrm{V}_{\mathrm{IN}}=12 \mathrm{~V}, \mathrm{Ta}=25^{\circ} \mathrm{C}$ )

| Symbol | Item | Conditions | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| In | VIN Consumption Current | $\mathrm{V}_{\mathrm{IN}}=30 \mathrm{~V}, \mathrm{~V}_{\mathrm{FB}}=1.0 \mathrm{~V}$ |  | 0.5 | 1.0 | mA |
| Vuvlo1 | UVLO Detect Voltage | Falling | 3.6 | 3.8 | 4.0 | V |
| Vuvloz | UVLO Released Voltage | Rising |  | $\begin{gathered} \hline \text { VUVLO1 } \\ +0.2 \end{gathered}$ | 4.2 | V |
| $V_{\text {FB }}$ | VFB Voltage Tolerance |  | 0.788 | 0.800 | 0.812 | V |
| $\Delta \mathrm{V}_{\mathrm{FB}} / \Delta \mathrm{Ta}$ | VFB Voltage Temperature Coefficient | $-40^{\circ} \mathrm{C} \leq \mathrm{Ta} \leq 85^{\circ} \mathrm{C}$ |  | $\pm 150$ |  | ppm/ ${ }^{\circ} \mathrm{C}$ |
| fosc | Oscillator Frequency |  | 1000 | 1250 | 1500 | kHz |
| $V_{\text {FLb }}$ | Fold-back Frequency (Ver. B) | $\mathrm{V}_{\mathrm{FB}}<0.56 \mathrm{~V}$ |  | 310 |  | kHz |
| Maxduty | Oscillator Max. Duty Cycle |  | 75 | 85 | 90 | \% |
| tmin | Minimum On Time |  |  | 100 |  | nsec |
| tss | Soft-start Time | $\mathrm{V}_{\mathrm{FB}}=0.72 \mathrm{~V}$ | 0.2 | 0.4 | 0.6 | ms |
| tdly | Delay Time for Latch Protection (Ver. A) |  | 1 | 2 | 4 | ms |
| R LxH | Lx High Side Switch ON Resistance |  |  | 0.35 |  | $\Omega$ |
| ILXHOFF | Lx High Side Switch Leakage Current |  |  | 0 | 5 | $\mu \mathrm{A}$ |
| ILImLxh | Lx High Side Switch Limited Current |  |  | 2.0 |  | A |
| Vcel | CE "L" Input Voltage |  |  |  | 0.3 | V |
| $V_{\text {ceh }}$ | CE "H" Input Voltage |  | 1.6 |  |  | V |
| $\mathrm{I}_{\text {FB }}$ | VFB Input Current |  | -1.0 |  | 1.0 | $\mu \mathrm{A}$ |
| Icel | CE "L" Input Current |  | -1.0 |  | 1.0 | $\mu \mathrm{A}$ |
| Iceh | CE "H" Input Current |  | -1.0 |  | 1.0 | $\mu \mathrm{A}$ |
| TTSD | Thermal Shutdown Detect Temperature | Hysteresis $30^{\circ} \mathrm{C}$ |  | 160 |  | ${ }^{\circ} \mathrm{C}$ |
| Istandby | Standby Current | $\mathrm{V}_{\text {IN }}=30 \mathrm{~V}$ |  | 0 | 5 | $\mu \mathrm{A}$ |

## OPERATING DESCRIPTIONS

## OPERATION OF STEP-DOWN DCIDC CONVERTER AND OUTPUT CURRENT

The step-down DC/DC converter charges energy in the inductor ( L ) when the LX transistor turns on, and discharges the energy from the inductor when LX transistor turns off and controls with less energy loss, so that a lower output voltage (Vout) than the input voltage ( $\mathrm{V}_{\mathrm{IN}}$ ) can be obtained. The operation of the step-down DC/DC converter is explained in the following figures.


Step1. The Nch transistor turns on and the inductor current (i1) flows, $L$ is charged with energy. At this moment, i1 increases from the minimum inductor current (ILmin), which is 0 A , and reaches the maximum inductor current (ILmax) in proportion to the on-time period (ton) of the Nch transistor.

Step2. When the Nch transistor turns off, $L$ tries to maintain IL at ILmax, so $L$ turns the diode on and the inductor current (i2) flows into L.

Step3. i2 decreases gradually and reaches ILmin after the open-time period (topen) of the Nch transistor, and then the diode turns off. This is called discontinuous current mode.

As the output current (lout) increases, the off-time period (toff) of the Nch transistor runs out before IL reaches ILmin. The next cycle starts, and the Nch transistor turns on and the diode turns off, which means IL starts increasing from ILmin. This is called continuous current mode.

In the case of PWM mode, Vout is maintained by controlling ton. During PWM mode, the oscillator frequency (fosc) is being maintained constant.

## APPLICATION INFORMATION

## TYPICAL APPLICATION CIRCUIT



R1240x Typical Application Circuit

## External Parts

| $\mathrm{C}_{\mathrm{IN}}$ | $10 \mu \mathrm{~F}$, KTS500B106M55N0T00 (Nippon Chemi-Con) |
| :---: | :--- |
| Cout | $10 \mu \mathrm{~F}$, GRM31CR71E106K (Murata) |
| $\mathrm{C}_{\text {bSt }}$ | $0.1 \mu \mathrm{~F}$, GRM21BB11H104KA01L (Murata) |
| L | $4.7 \mu \mathrm{H}$, SLF7045T-4R7M2R0-PF (TDK) |
| D | CMS11 (Toshiba) |

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## OUTPUT CURRENT AND SELECTION OF EXTERNAL COMPONENTS

The following equations explain the relationship between output current and peripheral components.

Ripple Current P-P value is described as IRP, ON resistance of switch is described as RonP, forward drop voltage is described as $V_{F}$, and DC resistance of inductor is described as RL.

First, when the switch is turned on, the following equation is satisfied.
$\mathrm{V}_{\text {IN }}=\mathrm{V}_{\text {OUT }}+\left(\right.$ RONH $\left.+\mathrm{R}_{\mathrm{L}}\right) \times$ lout $+\mathrm{L} \times \mathrm{I}_{\text {RP }} /$ ton
Equation 1

Second, when the switch is turned off, the diode is turned on, the following equation is satisfied.
$\mathrm{L} \times \mathrm{I}_{\mathrm{RP}} /$ toff $=\mathrm{V}_{\mathrm{F}}+\mathrm{V}_{\text {OUt }}+\mathrm{R}_{\mathrm{L}} \times$ lout
Equation 2

Put Equation 2 into Equation 1 to solve the ON duty of the switch (Don $=$ ton $/($ toff + ton $)$ ):
$D_{\text {ON }}=\left(V_{\text {OUT }}+V_{F}+R_{L} \times\right.$ IOUT $) /\left(V_{\text {IN }}+V_{F}-R_{\text {ONH }} \times\right.$ IOUT $)$
Equation 3

Ripple Current is described as follows:

Equation 4

Peak current that flows through $L$ and the switch is described as follows:

ILmax $=$ lout $+\mathrm{I}_{\mathrm{RP}} / 2$
Equation 5

Notes: Please consider ILmax when setting conditions of input and output, as well as selecting the external components. The above calculation formulas are based on the ideal operation of the ICs in continuous mode.

## TECHNICAL NOTES

The performance of a power source circuit using this device is highly dependent on a peripheral circuit. A peripheral component or the device mounted on PCB should not exceed its voltage, current or power ratings. When designing a peripheral circuit, please be fully aware of the following points. (Refer to our PCB layout for detailed information).

- External components must be connected as close as possible to the ICs and make wiring as short as possible. Especially, the capacitor connected in between VIN and GND pin must be wiring the shortest. The operating may be unstable due to the change of the electric potential of internal ICs by the switching current when the impedance of the power supply line and GND line is high. Make the power supply and GND lines sufficient. It is also necessary to give careful consideration to design the wiring of the power supply, GND, Lx, VOUT and the inductor because of the large current by the function of switching is flowing into them. Besides, the wiring between the resistance (R1), which set the output voltage, and the wiring of the inductor must separate from the load wiring.
- The ceramic capacitors have low ESR (Equivalent Series Resistance) type are recommended for the ICs. The recommendation of $\mathrm{C}_{\mathrm{IN}}$ capacitor between VIN and GND is more than $10 \mu \mathrm{~F}$, and Cout capacitor is more than $10 \mu \mathrm{~F}$ in the case $\mathrm{V}_{\text {out }} \geq 1.8 \mathrm{~V}$ or more than $20 \mu \mathrm{~F}$ in the case $1.8 \mathrm{~V}>\mathrm{V}$ out. Please check the bias dependence and the temperature variations of the ceramic capacitors.
- Normally, please select the inductor value in the range between $4.7 \mu \mathrm{H}$ and $10 \mu \mathrm{H}$ in the case of Vout $\geq$ $5 \mathrm{~V}, 4.7 \mu \mathrm{H}$ in the case of $5 \mathrm{~V}>\mathrm{V}_{\text {out }} \geq 1.8 \mathrm{~V}$ and $2.2 \mu \mathrm{H}$ in the case of $1.8 \mathrm{~V}>\mathrm{V}$ out. The internal phase compensation of this IC is designed with the above-mentioned inductor value and Cout ceramic capacitor value. When the inductor value is small, there is a possibility to trigger the over-current protection circuit by the peak switching current. As the peak switching current might reach to the limited value when the load current increase a lot.
- Please note; the over-current protection circuit is influenced by the temperature shift caused by operation of the IC.
- For the diode, please use the Schottky diode, which parasitic capacitance is small as possible, as, there is a possibility that the operating of IC becomes unstable by the large switching current.
- Output voltage is set by $\mathrm{V}_{\text {out }}=\mathrm{V}_{\mathrm{FB}} \times(\mathrm{R} 1+\mathrm{R} 2) / \mathrm{R} 2$. If the values of R 1 and R 2 are large, the impedance of VFB pin increases, and pickup the noise may result. The recommendation value range of $R 2$ is approximately between $1.2 \mathrm{k} \Omega$ to $16 \mathrm{k} \Omega$. If the operation may be unstable, reduce the impedance of VFB pin.


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Recommended Value for Each Output Voltage

| $\operatorname{Vout}(\mathrm{V})$ | 0.8 | 1 | 1.2 | 1.3 | 1.5 | $1.8 \sim 6$ | $6 \sim 15$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{R} 1(\mathrm{k} \Omega)$ | 0 | $=(\mathrm{Vout} / 0.8-1) \times 1.2$ |  |  |  |  |  |
| $\mathrm{R} 2(\mathrm{k} \Omega)$ | open | 1.20 | 1.20 | 1.20 | 1.20 | 1.20 | 1.20 |
| $\mathrm{C}_{\text {spd }}(\mathrm{pF})$ | open | 3300 | 2200 | 1500 | 470 | 470 | 330 |
| $\operatorname{Cout~}^{(\mu \mathrm{F})}$ ) | $22 \times 2$ | $10 \times 2$ | $10 \times 2$ | $10 \times 2$ | $10 \times 2$ | 10 | 10 |
| $\mathrm{~L}(\mu \mathrm{H})$ | 2.2 | 2.2 | 2.2 | 2.2 | 2.2 | 4.7 | $10.0(4.7)$ |

Recommended External Components

| Symbol | Condition | Value | Parts Name | MFR |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{Clin}^{\text {a }}$ |  | $\begin{aligned} & 10 \mu \mathrm{~F} / 50 \mathrm{~V} \\ & 10 \mu \mathrm{~F} / 50 \mathrm{~V} \\ & 10 \mu \mathrm{~F} / 50 \mathrm{~V} \end{aligned}$ | UMK325BJ106MM-P CGA6P3X7S1H106K KTS500B106M55NOT00 | TAIYO YUDEN TDK Nippon Chemi-Con |
| Cout | $\mathrm{V}_{\text {OUt }}>10 \mathrm{~V}$ $\begin{gathered} 10 \mathrm{~V}>\mathrm{V} \text { OUt }^{>} \mathrm{l} 1.8 \mathrm{~V} \\ \mathrm{~V}_{\text {out }}<1.8 \mathrm{~V} \end{gathered}$ | $\begin{aligned} & 10 \mu \mathrm{~F} / 50 \mathrm{~V} \\ & 10 \mu \mathrm{~F} / 50 \mathrm{~V} \\ & 10 \mu \mathrm{~F} / 50 \mathrm{~V} \\ & 10 \mu \mathrm{~F} / 25 \mathrm{~V} \\ & 22 \mu \mathrm{~F} / 10 \mathrm{~V} \end{aligned}$ | UMK325BJ106MM-P <br> CGA6P3X7S1H106K <br> KTS500B106M55NOT00 <br> GRM31CR71E106K <br> GRM31CR71A226M <br> NOTE: The value of Cout depends on the setting output voltage. | TAIYO YUDEN <br> TDK <br> Nippon Chemi-Con <br> Murata <br> Murata |
| $\mathrm{C}_{\text {bSt }}$ |  | $0.1 \mu \mathrm{~F} / 50 \mathrm{~V}$ | GRM21BB11H104KA01L | Murata |
| R ${ }_{\text {bSt }}$ |  | $51.0 \Omega$ |  |  |
| L | $40 \mathrm{~V} / 2.0 \mathrm{~A}$ | $\begin{aligned} & 10 \mu \mathrm{H} \\ & 4.7 \mu \mathrm{H} \\ & 2.2 \mu \mathrm{H} \end{aligned}$ | SLF6045T-100M1R6-3PF <br> SLF7045T-4R7M2R0-PF <br> VLCF4020T-2R2N1R7 | $\begin{aligned} & \hline \text { TDK } \\ & \text { TDK } \\ & \text { TDK } \end{aligned}$ |
| D | $\begin{aligned} & 30 \mathrm{~V} / 2.0 \mathrm{~A} \\ & 40 \mathrm{~V} / 2.0 \mathrm{~A} \end{aligned}$ | $\begin{aligned} & 0.32 \mathrm{~V} \\ & 0.49 \mathrm{~V} \end{aligned}$ | CMS06 <br> CMS11 <br> NOTE: Diode depends on the input voltage and output Current. | TOSHIBA TOSHIBA |
| Rce | The diode is connected between the CE pin and the VIN pin as the ESD protection element. If there is the possibility that the voltage of the CE pin becomes higher than the voltage of the VIN pin, it is recommended to connect the $5 \mathrm{k} \Omega$ resistance with the CE pin for preventing a large current flows into the VIN pin from the CE pin. |  |  |  |

## THE NOTE OF LAYOUT PATTERN

1. The wire of Power line ( $\mathrm{V}_{\mathrm{IN}}, \mathrm{GND}$ ) should be broad to minimize the parasitic inductance.

The Bypass capacitor must be connected as close as possible in between ViN - GND
2. The wire between Lx pin and the inductor as short as possible to minimize the parasitic inductance (This Evaluation Board is designed for the product evaluation board. Therefore large inductors or diodes can be set and the large space of $L x$ area has been secured.)
3. The ripple current flows through the output capacitor. If the GND side of the output capacitor is connected very close to GND pin of the IC, the noise might have a bad impact on the IC. Therefore, the GND side of the output capacitor is better to connect to the outside of the GND of the $\mathrm{CiN}_{\mathrm{IN}}$, or connect to the GND plain layer.
4. R1, R2, Cspd and Rspd should be mounted on the position as close as possible to the FB pin, and away from the inductor and BST pin.
5. The feed-back must be made as close as possible from the Output capacitor (Cout)

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## PCB LAYOUT

Evaluation board of R1240N001x


Evaluation board of R1240K003x


TYPICAL CHARACTERISTICS

1) Output Voltage VS. Output Current R1240x00Xx

2) Output Voltage VS. Input Voltage R1240x00Xx

3) Efficiency VS. output Current

R1240x00Xx


R1240x00Xx


R1240x00Xx


R1240x00Xx

4) FB Voltage VS. Temperature R1240x00Xx

6) Maxduty VS. Temperature R1240x00Xx

5) Oscillator Frequency VS. Temperature R1240x00Xx

7) Fold-Back Frequency VS. Temperature R1240x00XB


The power dissipation of the package is dependent on PCB material, layout, and environmental conditions. The following conditions are used in this measurement.

Measurement Conditions

|  | Standard Test Land Pattern |
| :---: | :---: |
| Environment | Mounting on Board (Wind Velocity $=0 \mathrm{~m} / \mathrm{s}$ ) |
| Board Material | Glass Cloth Epoxy Plastic (Double-Sided Board) |
| Board Dimensions | $40 \mathrm{~mm} \times 40 \mathrm{~mm} \times 1.6 \mathrm{~mm}$ |
| Copper Ratio | Top Side: Approx. $50 \%$ |
| Bottom Side: Approx. $50 \%$ |  |
| Through-holes | $\phi 0.5 \mathrm{~mm} \times 44 \mathrm{pcs}$ |

Measurement Result
$\left(\mathrm{Ta}=25^{\circ} \mathrm{C}, \mathrm{Tjmax}=125^{\circ} \mathrm{C}\right)$

|  | Standard Test Land Pattern |
| :---: | :---: |
| Power Dissipation | 430 mW |
| Thermal Resistance | $\theta \mathrm{ja}=\left(125-25^{\circ} \mathrm{C}\right) / 0.43 \mathrm{~W}=233^{\circ} \mathrm{C} / \mathrm{W}$ |



Power Dissipation vs. Ambient Temperature


IC Mount Area (mm)

Measurement Board Pattern


UNIT: mm

SOT-23-6W Package Dimensions (Unit: mm)

The power dissipation of the package is dependent on PCB material, layout, and environmental conditions. The following conditions are used in this measurement.

## Measurement Conditions

|  | High Wattage Land Pattern | Standard Land Pattern |
| :---: | :---: | :---: |
| Environment | Mounting on Board (Wind Velocity $=0 \mathrm{~m} / \mathrm{s}$ ) | Mounting on Board (Wind Velocity $=0 \mathrm{~m} / \mathrm{s}$ ) |
| Board Material | Glass Cloth Epoxy Plastic (Four-Layer Board) | Glass Cloth Epoxy Plastic (Double-Sided Board) |
| Board Dimensions | $35 \mathrm{~mm} \times 90 \mathrm{~mm} \times 0.8 \mathrm{~mm}$ | $40 \mathrm{~mm} \times 40 \mathrm{~mm} \times 1.6 \mathrm{~mm}$ |
| Copper Ratio | Outer Layers (First and Fourth Layers): Approx.15\% Inner Layers (Second and Third Layers): Approx.15\% | Top Side: Approx. 50\% Bottom Side: Approx. 50\% |
| Copper Foil Thickness | Outer Layers (First and Fourth Layers): Approx. $35 \mu \mathrm{~m}$ Inner Layers (Second and Third Layers): Approx. $18 \mu \mathrm{~m}$ | Top Side: Approx. $35 \mu \mathrm{~m}$ Bottom Side: Approx. $35 \mu \mathrm{~m}$ |
| Through-holes | $\phi 0.3 \mathrm{~mm} \times 9$ holes (connecting outer and inner layers to a package tab) <br> $\phi 0.5 \mathrm{~mm} \times 10$ holes (connecting pins) | $\phi 0.54 \mathrm{~mm} \times 30$ holes |

Measurement Result
$\left(\mathrm{Ta}=25^{\circ} \mathrm{C}, \mathrm{Tjmax}=125^{\circ} \mathrm{C}\right)$

|  | High Wattage Land Pattern | Standard Land Pattern |
| :---: | :---: | :---: |
| Power Dissipation | $1400 \mathrm{~mW}\left(\mathrm{Tjmax}=125^{\circ} \mathrm{C}\right)$ | $910 \mathrm{~mW}\left(\mathrm{Tjmax}=125^{\circ} \mathrm{C}\right)$ |
| Thermal Resistance | $\theta \mathrm{ja}=\left(125-25^{\circ} \mathrm{C}\right) / 1.4 \mathrm{~W}=71^{\circ} \mathrm{C} / \mathrm{W}$ | $\theta \mathrm{jc}=\left(125-25^{\circ} \mathrm{C}\right) / 0.91 \mathrm{~W}=110^{\circ} \mathrm{C} / \mathrm{W}$ |




High Wattage

() IC Mount Area (mm)

Measurement Board Pattern


DFN(PLP)2527-10 Package Dimensions

* The tab on the bottom of the package is substrate level (GND). It is recommended that the tab be connected to the ground plane on the board, or otherwise be left floating.

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