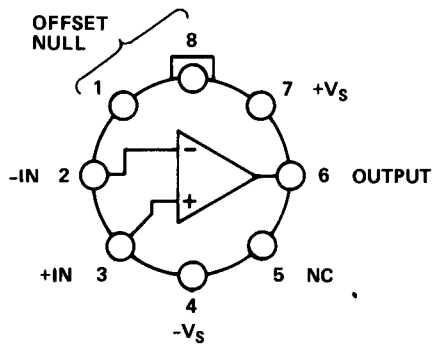
Low Cost, Laser Trimmed, Precision IC Op Amp

AD517

FEATURES

Low Input Bias Current: 1nA max (AD517L)
Low Input Offset Current: 0.25nA max (AD517L)
Low V_{OS}: 50μV max (AD517L), 150μV max (AD517J)
Low V_{OS} Drift: 1.3μV/°C (AD517L)
Internal Compensation
MIL-Standard Parts Available
8-Pin TO-99 Hermetic Metal Can
Available in Chip Form

PIN CONFIGURATION



TOP VIEW

PRODUCT DESCRIPTION

The AD517 is a high accuracy monolithic op amp featuring extremely low offset voltages and input currents. Analog Devices' thermally-balanced layout and superior IC processing combine to produce a truly precision device at low cost.

The AD517 is laser trimmed at the wafer level (LWT) to produce offset voltages less than 50μV and offset voltage drifts less than 1.3μV/°C unnulled. Superbeta input transistors provide extremely low input bias currents of 1nA max and offset currents as low as 0.25nA max. While these figures are comparable to presently available BIFET amplifiers at room temperature, the AD517 input currents decrease, rather than increase, at elevated temperatures. Open-loop gain in many IC amplifiers is degraded under loaded conditions due to thermal gradients on the chip. However, the AD517 layout is balanced along a thermal axis, maintaining open-loop gain in excess of 1,000,000 for a wide range of load resistances.

The input stage of the AD517 is fully protected, allowing differential input voltages of up to $\pm V_S$ without degradation of gain or bias current due to reverse breakdown. The output stage is short-circuit protected and is capable of driving a load capacitance up to 1000pF.

The AD517 is well suited to applications requiring high precision and excellent long-term stability at low cost, such as stable references, followers, bridge instruments and analog computation circuits.

The circuit is packaged in a hermetically sealed TO-99 metal can, and is available in three performance versions (J, K, and L) specified over the commercial 0 to +70°C range; and one version (AD517S) specified over the extended temperature range, -55°C to +125°C.

PRODUCT HIGHLIGHTS

1. Offset voltage is 100% tested and guaranteed on all models.
2. The AD517 exhibits extremely low input bias currents without sacrificing CMRR (over 100dB) or offset voltage stability.
3. The AD517 inputs are protected (to $\pm V_S$), preventing offset voltage and bias current degradation due to reverse breakdown of the input transistors.
4. Internal compensation is provided, eliminating the need for additional components (often required by high accuracy IC op amps).
5. The AD517 can directly replace 725, 108, and AD510 amplifiers. In addition, it can replace 741-type amplifiers if the offset-nulling potentiometer is removed.
6. Thermally-balanced layout insures high open-loop gain independent of thermal gradients induced by output loading, offset nulling, and power supply variations.
7. Chips are available.

AD517—SPECIFICATIONS (@ +25°C and V_S = ±15V dc)

Model	AD517J Min	AD517J Typ	AD517J Max	AD517K Min	AD517K Typ	AD517K Max	AD517L Min	AD517L Typ	AD517L Max	AD517S Min	AD517S Typ	AD517S Max	Units
OPEN LOOP GAIN													
V_O = ±10V, $R_L \geq 2k\Omega$	**10^6**			**10^6**			**10^6**			**10^6**			V/V
T_{min} to T_{max}, $R_L = 2k\Omega$	500,000			500,000			500,000			**250,000**			V/V
OUTPUT CHARACTERISTICS													
Voltage @ $R_L = 2k\Omega$, T_{min} to T_{max}	**±10**			**±10**			**±10**			**±10**			V
Load Capacitance		1000			1000			1000			1000		pF
Output Current	10			10			10			10			mA
Short Circuit Current		25			25			25			25		mA
FREQUENCY RESPONSE													
Unity Gain Small Signal		250			250			250			250		kHz
Full Power Response		1.5			1.5			1.5			1.5		kHz
Slew Rate, Unity Gain		0.10			0.10			0.10			0.10		V/µs
INPUT OFFSET VOLTAGE													
Initial Offset			**150**			**75**			**50**			**75**	µV
Input Offset vs. Temp.			3.0			**1.8**			**1.3**			**1.8**	µV/°C
Input Offset vs. Supply			**25**			**10**			**10**			**10**	µV/V
T_{min} to T_{max}			40			15			15			**20**	µV/V
INPUT BIAS CURRENT													
Initial			**5**			**2**			**1.0**			**2.0**	nA
T_{min} to T_{max}			8			3.5			1.5			**10**	nA
vs. Temp, T_{min} to T_{max}			±20			±10			±4			±10	pA/°C
INPUT OFFSET CURRENT													
Initial			**1.0**			**0.75**			**0.25**			**2.0**	nA
T_{min} to T_{max}			1.5			1.25			0.4			**10**	nA
INPUT IMPEDANCE													
Differential		15‖1.5			20‖1.5			20‖1.5			20‖1.5		MΩ‖pF
Common Mode		2.0×10^{9}			2.0×10^{9}			2.0×10^{9}			2.0×10^{9}		Ω
INPUT VOLTAGE RANGE													
Differential		$\pm V_S$			$\pm V_S$			$\pm V_S$			$\pm V_S$		V
Common Mode Rejection	**94**			**110**			**110**			**110**			dB
Common Mode Rejection T_{min} to T_{max}	94			110			100			**100**			dB
INPUT NOISE													
Voltage, 0.1Hz to 10Hz		2			2			2			2		µV p-p
f = 10Hz		35			35			35			35		nV/√Hz
f = 100Hz		25			25			25			25		nV/√Hz
f = 1kHz		20			20			20			20		nV/√Hz
Current, f = 10kHz		0.05			0.05			0.05			0.05		pA/√Hz
f = 100Hz		0.03			0.03			0.03			0.03		pA/√Hz
f = 1kHz		0.03			0.03			0.03			0.03		pA/√Hz
POWER SUPPLY													
Rated Performance		±15			±15			±15			±15		V
Operating	±5		±18	±5		±18	±5		±18	±5		±18	V
Quiescent Current			**4**			**3**			**3**			**3**	mA
TEMPERATURE RANGE													
Operating, Rated Performance	0		+70	0		+70	0		+70	−55		+125	°C
Storage	−65		+150	−65		+150	−65		+150	−65		+150	°C
PACKAGE OPTION[1]													
TO-99 Style (H-08B)		AD517JH			AD517KH			AD517LH			AD517SH		
J and S Grade Chips Also Available													

NOTES

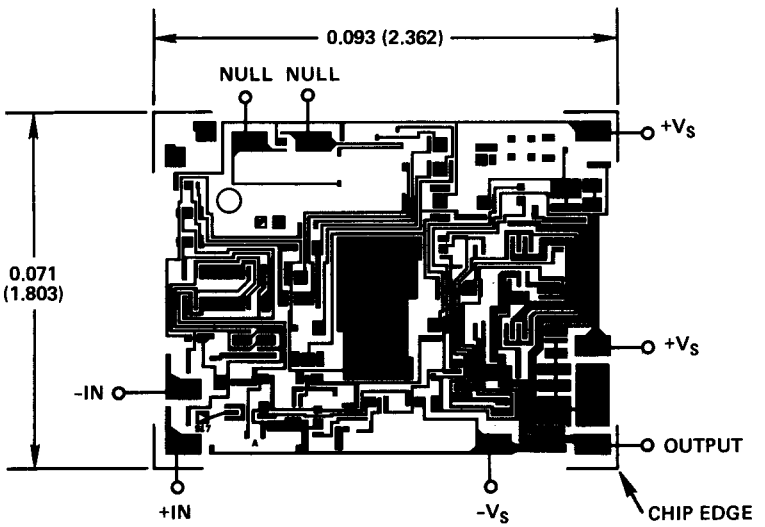
[1]For outline information see Package Information section.

Specifications subject to change without notice.

Specifications shown in boldface are tested on all production units at final electrical test. Results from those tests are used to calculate outgoing quality levels. All min and max specifications are guaranteed, although only those shown in boldface are tested on all production units.

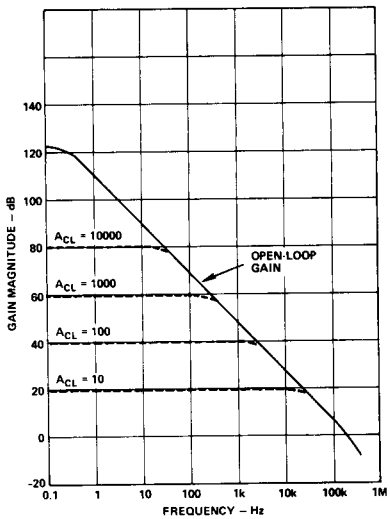
CHIP DIMENSIONS AND BONDING DIAGRAM

Contact factory for latest dimensions.
Dimensions shown in inches and (mm).

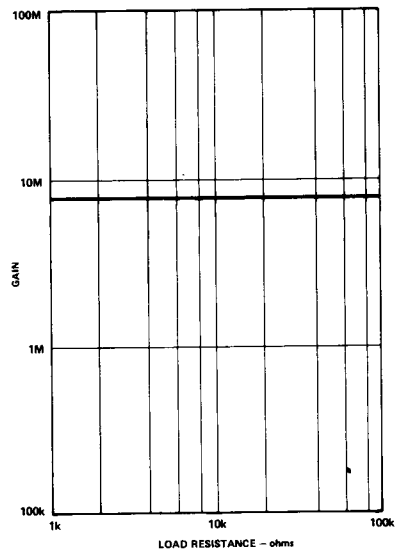


THE AD517 IS AVAILABLE IN
LASER-TRIMMED CHIP FORM.

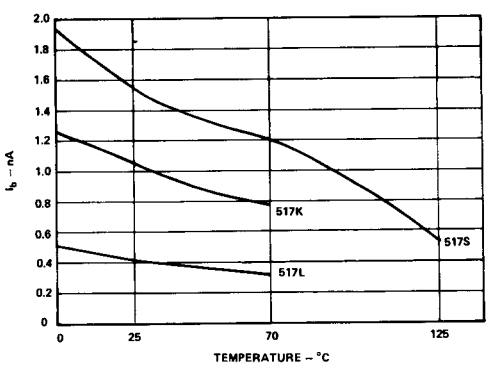
AD517—Typical Performance Curves



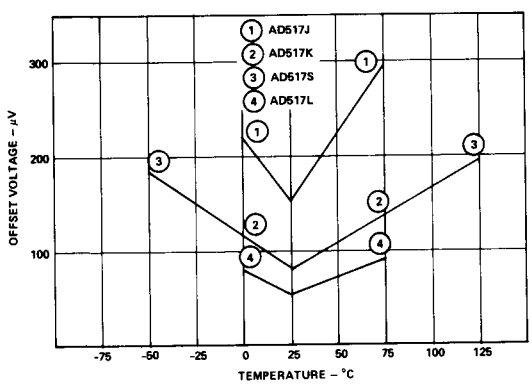
Small-Signal Gain vs. Frequency



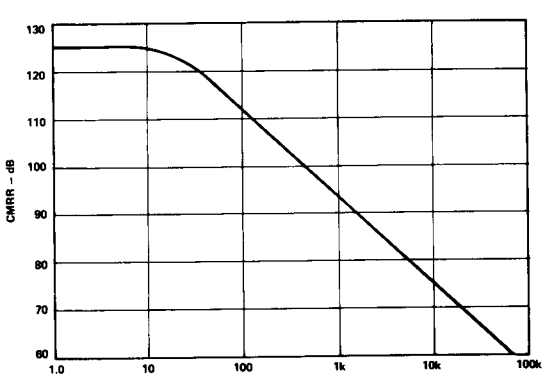
Open-Loop Gain vs. Load Resistance



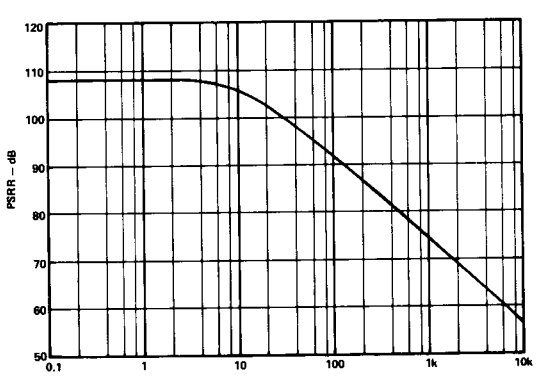
Input Bias Current vs. Temperature



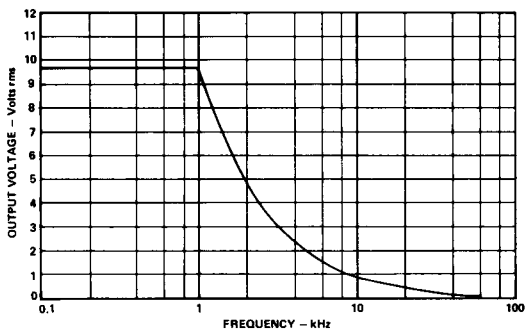
Untrimmed Offset Voltage vs. Temperature



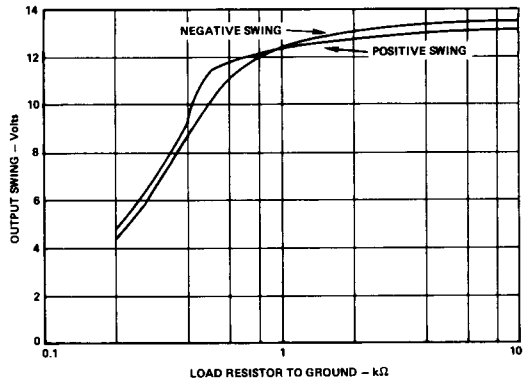
CMRR vs. Frequency



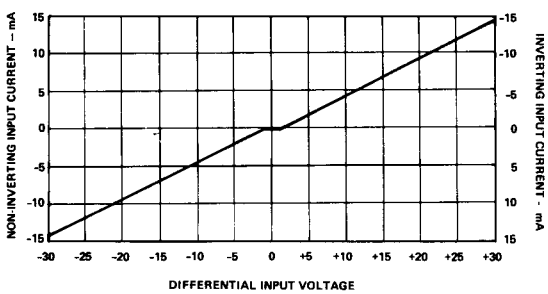
PSRR vs. Frequency



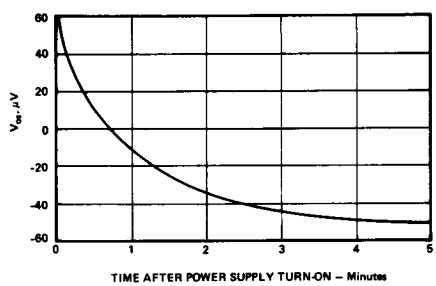
Maximum Undistorted Output vs. Frequency (Distortion ≤ 1%)



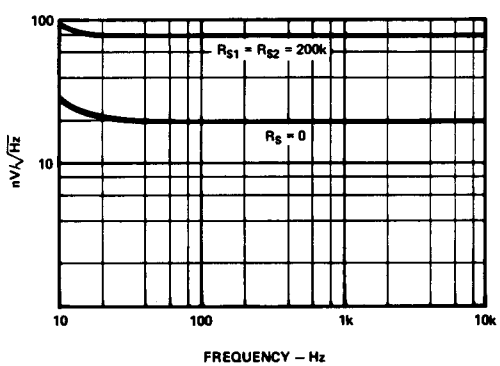
Output Voltage vs. Load Resistance



Input Current vs. Differential Input Voltage

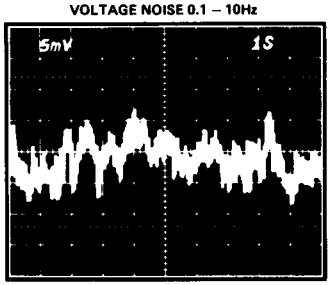


Warm-Up Offset Voltage Drift



Total Input Noise Voltage vs. Frequency

VOLTAGE NOISE 0.1 – 10Hz



E_n = 500nV/DIV RTI

Low Frequency Voltage Noise (0.1 to 10Hz)

AD517—Applications

NULLING THE AD517

The internally-trimmed offset voltage of the AD517 will be low enough for most circuits without further nulling. However, in high precision applications, the AD517 may be nulled using either of the following methods:

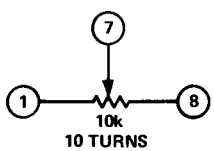
Figure 1A shows a simple circuit using a 10kΩ, ten-turn potentiometer. This circuit allows nulling to within several microvolts.

The circuit of Figure 1B is recommended in applications where nulling to within 1μV is desired. This circuit has the advantage that potentiometer instability effects are reduced by a factor of ten. Values of R_1' and R_2' are calculated as follows:

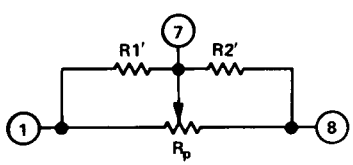
1. Null the offset to zero using a standard 10k pot, as shown in Figure 1A.
2. Measure pot halves R_1 and R_2.
3. Calculate:

$$R_1' = \frac{R_1 \times 50k\Omega}{50k\Omega - R_1}, \qquad R_2' = \frac{R_2 \times 50k\Omega}{50k\Omega - R_2}$$

4. Replace the pot with R_1' and R_2' using the closest value 1% metal film resistors.
5. Use a 100k, ten-turn pot for R_p to complete the nulling.



A. Simple



B. High Precision

Figure 1. Nulling Circuits

AN INSTRUMENT INPUT AMPLIFIER USING THE AD517L

The circuit shown in Figure 2 represents a typical input stage for laboratory instruments and panel meters. The amplifier is non-inverting and offers selectable gains from 1 to 1000 in decade steps.

Input impedance of this amplifier is 10 megohms, determined by resistor R_1. The offset nulling network comprised of R_3, R_4 and R_5 is the same one described earlier. If a less precise adjustment can be tolerated, a single 10k potentiometer can be substituted for R_3, R_4 and R_5.

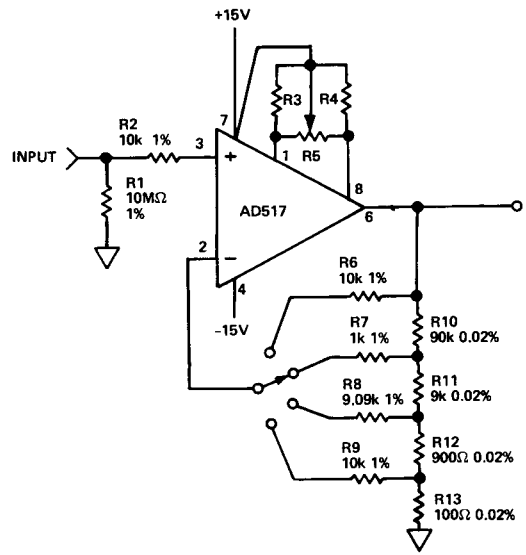


Figure 2. Stable Instrument Input Amplifier

Gain switching is accomplished in the feedback network. The divider consisting of R_{10}, R_{11}, R_{12} and R_{13} determines the gain by dividing the output and returning it to the inverting input of the amplifier. The ratio tolerances of these resistors uniquely determine the gain of the amplifier. The impedance seen by the inverting input is held constant at 10k ohms by R_6, R_7 R_8 or R_9 depending on the gain selected. Since input bias currents flow through equal resistances, the offset voltages produced will cancel each other. The input offset currents will produce an insignificant offset voltage on the order of 1 microvolt. If this offset is nulled out at the highest gain selected, it will be nulled on all ranges.

The AD517 offers excellent temperature stability in this circuit. Once the offset has been zeroed, the error produced by offset current drift will remain quite low due to the extremely low offset current drift of the AD517. A FET-input op amp would not work well in this application, since the input offset currents would double for each 10°C increase in temperature, soon exceeding the input offset currents of the AD517.